

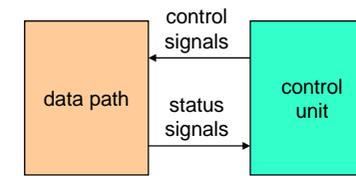
# Chapter 8

## 데이터 경로와 제어장치 및 산술연산 회로

## 8.1 데이터 경로와 제어장치

### 데이터경로와 제어장치

- 많은 순차 회로의 설계는 다음의 두 부분으로 구성
  - datapath: data의 이동 및 연산을 위한 장치 control unit에 상태신호 제공
  - control unit: datapath에서 적절한 순서로 data 이동 및 연산을 수행할 수 있도록 제어신호 제공.
- 먼저, datapath를 설계
- 다음에, control unit를 설계



## 8.2 수의 표현

### 2진수 표현: $B = b_{n-1}b_{n-2} \dots b_1b_0$

- unsigned integer:

$$\text{값} = \text{unsigned}(B) = \sum_{i=0}^{n-1} b_i 2^i$$

- signed integer (2's complement 표기법):

$$\text{값} = \text{signed}(B) = -b_{n-1} 2^{n-1} + \sum_{i=0}^{n-2} b_i 2^i$$

### (예) $B = 10100010$

- unsigned(B) =  $2^7 + 2^5 + 2^1 = 128 + 32 + 2 = 162$
- signed(B) =  $-2^7 + 2^5 + 2^1 = -128 + 32 + 2 = -94$  또는  $-2^8 + (2^7 + 2^5 + 2^1) = -256 + 162 = -94$

(참고) 10100010의 2의 보수

$$= 01011110 = 2^6 + 2^4 + 2^3 + 2^2 + 2^1 = 64 + 16 + 8 + 4 + 2 = 94$$

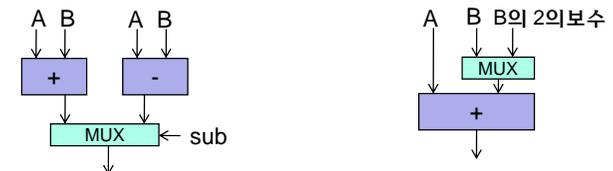
## 8.3 가감산기와 shifter

### 가산기 설계

- 방법1: 논리회로를 직접 설계
- 방법2: + 연산자 이용
- FPGA를 사용한 설계에서는 + 연산자를 이용하는 것이 좋음  
→ 합성도구가 carry 전용 경로를 사용하므로 더 효율적인 합성을 수행

### 가감산기 설계

- 가산기, 감산기 별도 구현
- 2의 보수 덧셈을 사용하여 뺄셈 수행



## 가산기 : 논리회로 직접 설계

```

module adder(data1, data2, cin, sum, cout);
  input [3:0] data1, data2;
  input cin;
  output [3:0] sum;
  output cout;

  integer i;
  reg [4:0] c;

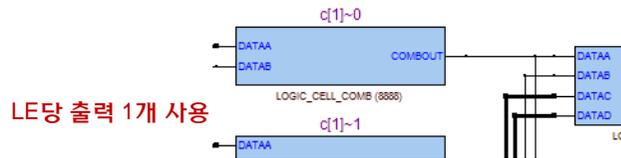
  assign sum = data1 ^ data2 ^ c[3:0];
  assign cout = c[4];

  always @* begin
    c[0] = cin;
    for (i=0; i<4; i=i+1) begin
      c[i+1] = data1[i] & data2[i] | data1[i] & c[i];
    end
  end
endmodule

```

### 합성결과

Total logic elements	10
Total combinational functions	10



LE당 출력 1개 사용

## 가산기 : + 연산자 이용

```

module adder(data1, data2, cin, sum, cout);
  input [3:0] data1, data2;
  input cin;
  output [3:0] sum;
  output cout;

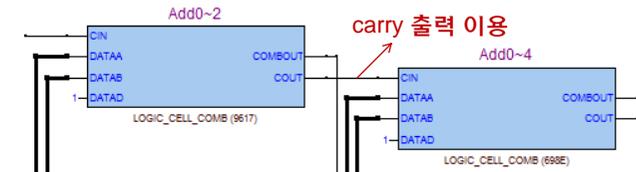
  assign {cout, sum} = data1 + data2 + cin;
endmodule

```

### 합성결과

Total logic elements	6
Total combinational functions	6

더 효율적

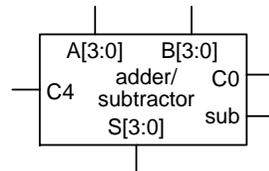


carry 출력 이용

## 가감산기

### 가감산기 동작

- sub=0일때 가산기
  - C0, C4는 carry 입출력
- sub=1일때 감산기
  - C0, C4는 borrow 입출력



### 가감산기 설계 - 가산기와 감산기 이용

```

module addsub(sub, a, b, cin, s, cout);
  input sub;
  input [3:0] a, b;
  input cin;
  output [3:0] s;
  output cout;

  assign {cout, s} = sub ? (a - b - cin) : (a + b + cin);
endmodule

```

가산기, 감산기 개별적 구현 가능성

## 가감산기(2)

### 가감산기 설계 - 뺄셈을 2의 보수 덧셈으로 구현

- 2의 보수 = (1의 보수) + 1
- borrow=0이면 A + (B의 1의 보수) + 1
- borrow=1이면 A + (B의 1의 보수) + 0
- borrow출력은 가산기의 carry출력과 반대가 됨

```

module addsub(sub, a, b, cin, s, cout);
  input sub;
  input [3:0] a, b;
  input cin;
  output [3:0] s;
  output cout;

  wire [3:0] b2;
  wire cin2, cout2;

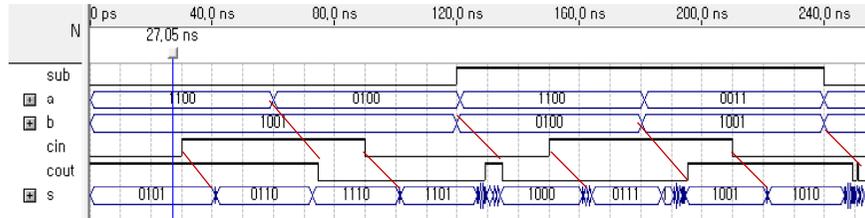
  assign b2 = sub ? ~b : b;
  assign cin2 = sub ? ~cin : cin;
  assign {cout2, s} = a + b2 + cin2;
  assign cout = sub ? ~cout2 : cout2;
endmodule

```

가산기

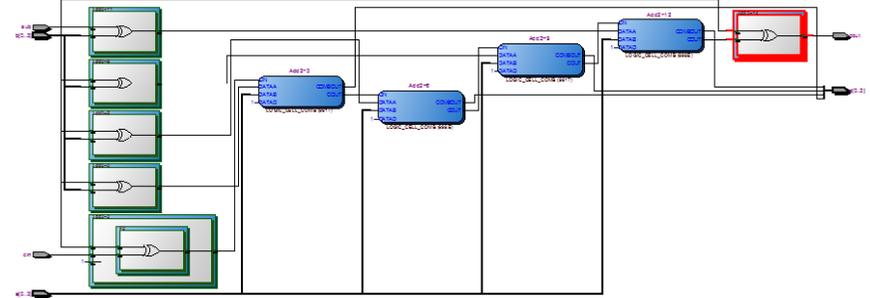
### 가감산기(3)

■ simulation (두 설계가 같음)



### 가감산기(4)

■ 합성결과

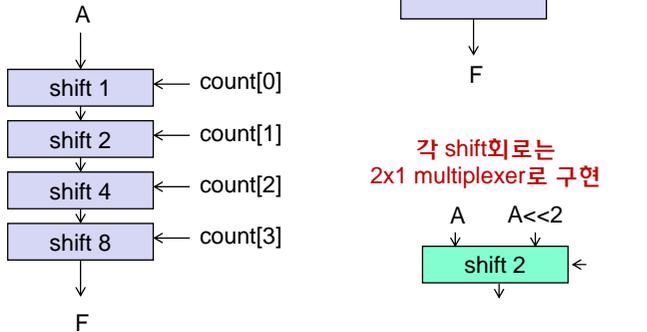


### Shifter

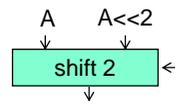
■ Barrel Shifter

- 한 번에 여러 비트(가변)를 이동시킬 수 있는 조합회로 (shift 횟수=count)

■ 구현



각 shift회로는 2x1 multiplexer로 구현



■ 시프트 연산자 사용:  $A \ll count \rightarrow$  합성도구가 구현

### 8.4 곱셈기

■ 2진수 곱셈 (unsigned)

```

      1101   Multiplicand (13)
    x) 1011   Multiplier(11)
    -----
      1101
     0000
    1101
   -----
  10001011   Product(143)
    
```

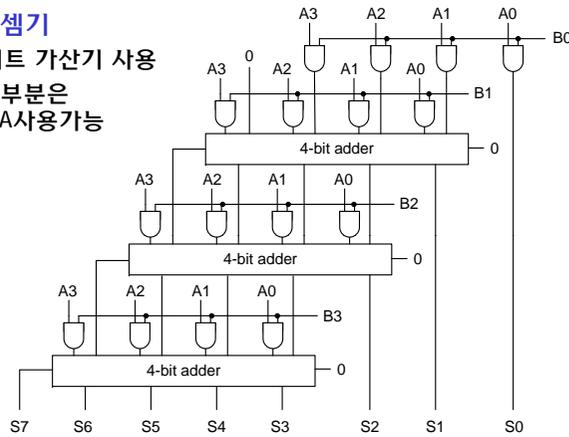
■ 곱셈기 설계

- 조합회로 설계: 배열 곱셈기
  - n-1개의 n-bit 가산기를 사용
- 순차회로 설계:
  - 1개의 n-bit 가산기를 사용
  - 한번에 한 자리씩 곱셈 수행  $\rightarrow$  n번 수행

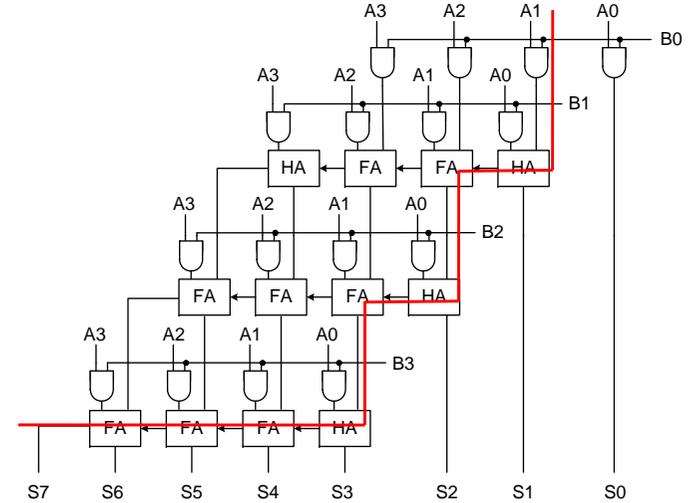
# 배열 곱셈기

## n비트 배열 곱셈기

- n-1개의 n비트 가산기 사용
- 0을 더하는 부분은 FA대신에 HA사용가능

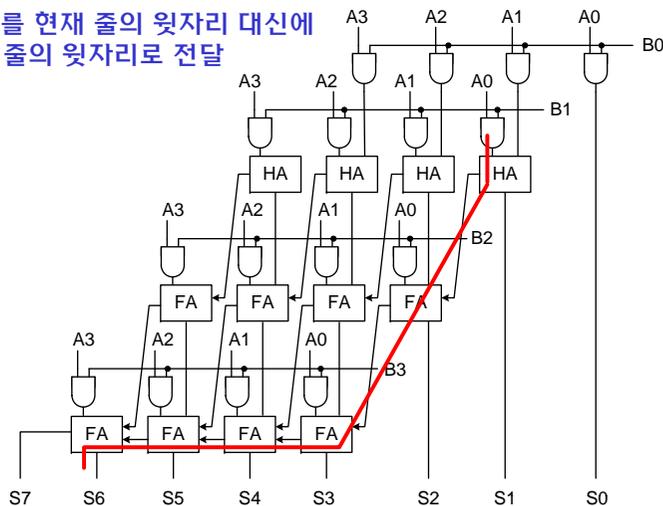


## 곱셈연산자 사용 : $A * B \rightarrow$ 합성도구가 조합회로 곱셈기 구현



# Carry Save 배열 곱셈기

- carry를 현재 줄의 윗자리 대신에 다음 줄의 윗자리로 전달



# 8.5 순차 곱셈기

## 2진수 곱셈 과정

```

1101 Multiplicand
x) 1011 Multiplier
-----
1101 ----- Shift Left Multiplicand
100111
 0000
100111
1101
10001111 Product
    
```

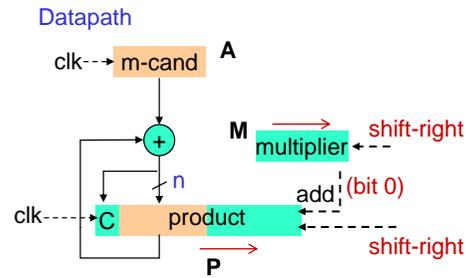
## 순차곱셈기

- 한 번에 multiplier 1자리씩 곱셈 계산
- datapath 와 control unit 로 구성됨
  - datapath를 설계하면서 필요한 제어신호 정의
  - control unit에서 적절한 순서로 제어신호를 공급

# Datapath

## Algorithm

$M = \text{multiplier (n bit)}$   
 $A = \text{multiplicand (n-bit)}$   
 $P = 0 \text{ (2n-bit)}$   
**repeat**  $n$  **times do**  
   if ( $M[0]=1$ )  $P_H = P_H + A$   
   **shift right**  $M$   
   **shift right**  $P$   
**end**



$P_H$ 는 P의 upper half

## 특징

- multiplicand를 shift-left하는 것 대신에 product를 shift-right함
- n-bit adder 및 n-bit multiplicand register 사용

```

module Datapath (product, m0, word1, word2, Load, Shift, Add, reset, clock);
  output [7:0] product;
  output m0;
  input [3:0] word1, word2;
  input Load, Shift, Add, clock, reset;
  reg [7:0] product;
  reg [3:0] multiplier, multiplicand;
  reg c_out;
  wire m0;

```

word1 : multiplicand  
word2 : multiplier

control unit에서 제어신호를 공급

```

assign m0 = multiplier[0];

always @ (posedge clock or posedge reset) begin
  if (reset) begin
    multiplicand <= 0; multiplier <= 0; product <= 0;
  end else if (Load) begin
    multiplicand <= word1; multiplier <= word2; product <= 0;
  end else if (Shift) begin
    multiplier <= multiplier >> 1;
    {c_out, product} <= {c_out, product} >> 1;
  end else if (Add) begin
    {c_out, product[7:4]} <= product[7:4] + multiplicand;
  end
end
endmodule

```

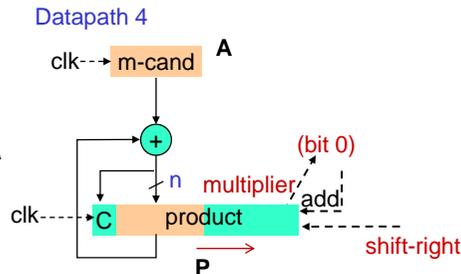
# Datapath (product와 multiplier를 같이 사용)

## 수정사항

- Product의 하위부분에 multiplier 저장  
→ multiplier 레지스터 불필요

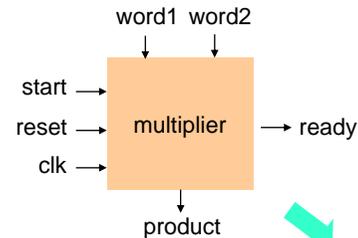
## Algorithm

$A = \text{multiplicand (n-bit)}$   
 $P_L = \text{multiplier}$   
 $P_H = 0 \text{ (P: 2n-bit)}$   
**repeat**  $n$  **times do**  
   if ( $M[0]=1$ )  $P_H = P_H + A$   
   **shift right**  $M$   
   **shift right**  $P$   
**end**



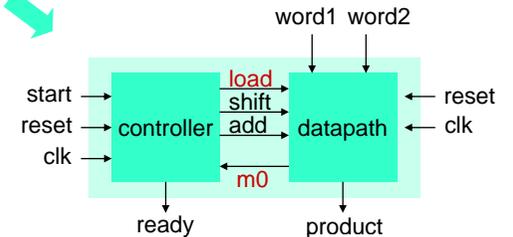
# 순차 곱셈기

## 블록 다이어그램과 내부 구조도



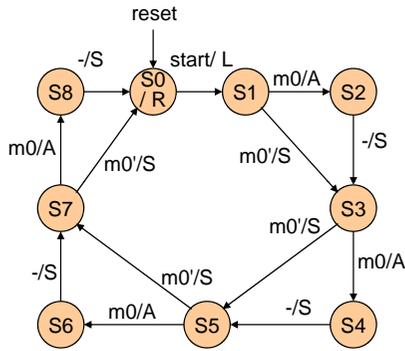
reset: 상태 초기화  
 start: 곱셈 수행 시작  
 ready: 계산할 준비됨

word1: 피승수  
 word2: 승수  
 product: 곱



# 순차 곱셈기

## controller의 상태도



L: Load  
R: Ready  
A: Add  
S: Shift

Load: 입력을 레지스터에 저장하는 제어신호

shift 또는 add-shift 동작을 반복

in/out 은  
in=1일 때 out=1이 됨을 표시  
in/out은  
in=0일 때 out=1이 됨을 표시  
표시되지 않은 출력은 0을 표시

# 순차곱셈기

## Verilog description

```

module multiplier1(product, Ready, word1, word2, Start, reset, clock, state);
    output [7:0] product;
    output Ready;
    input [3:0] word1, word2;
    input Start, clock, reset;
    output [3:0] state; // for observation
    wire m0, Load, Shift, Add;

    Datapath u1 (product, m0, word1, word2, Load, Shift, Add, reset, clock);
    Controller u2 (Load, Shift, Add, Ready, m0, Start, reset, clock, state);
endmodule
    
```

Datapath는 앞의 정의 참조



```

module Controller (Load, Shift, Add, Ready, m0, Start, reset, clock, state);
    output Load, Shift, Add, Ready;
    input m0, Start, clock, reset;
    output [3:0] state; // for observation
    reg [3:0] state, next_state;
    parameter S0 = 0, S1 = 8, S2 = 9, S3 = 10, S4 = 11,
              S5 = 12, S6 = 13, S7 = 14, S8 = 15;
    reg Load, Shift, Add;
    wire Ready;

    // State transition
    always @ (posedge clock or posedge reset) begin
        if (reset) state <= S0;
        else state <= next_state;
    end
    
```

초기상태와 곱셈이 진행되는 8개 상태를  
최상위비트로 구분하도록 상태 인코딩함

(계속)



```

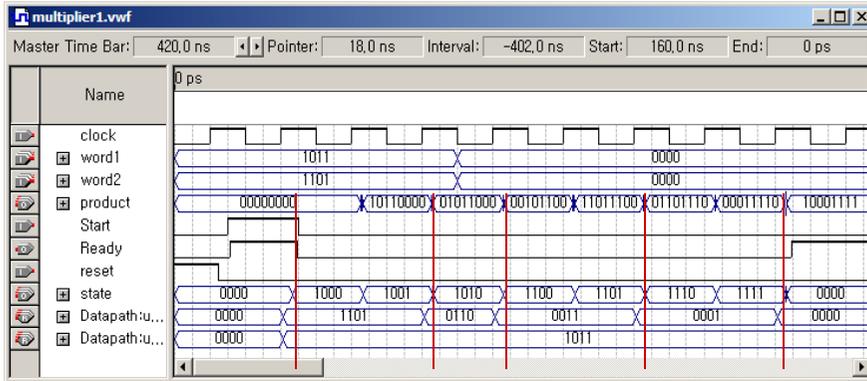
// Next state and control output logic
always @ (state or Start or m0) begin
    Load = 0; Shift = 0; Add = 0;
    case (state)
        S0: if (Start) begin Load = 1; next_state = S1; end
            else next_state = S0;
        S1: if (m0) begin Add = 1; next_state = S2; end
            else begin Shift = 1; next_state = S3; end
        S2: begin Shift = 1; next_state = S3; end
        S3: if (m0) begin Add = 1; next_state = S4; end
            else begin Shift = 1; next_state = S5; end
        S4: begin Shift = 1; next_state = S5; end
        S5: if (m0) begin Add = 1; next_state = S6; end
            else begin Shift = 1; next_state = S7; end
        S6: begin Shift = 1; next_state = S7; end
        S7: if (m0) begin Add = 1; next_state = S8; end
            else begin Shift = 1; next_state = S0; end
        S8: begin Shift=1; next_state = S0; end
        default: next_state = S0;
    endcase
end

assign Ready = (state==S0) && ~reset;
endmodule
    
```

# 순차 곱셈기

## Simulation

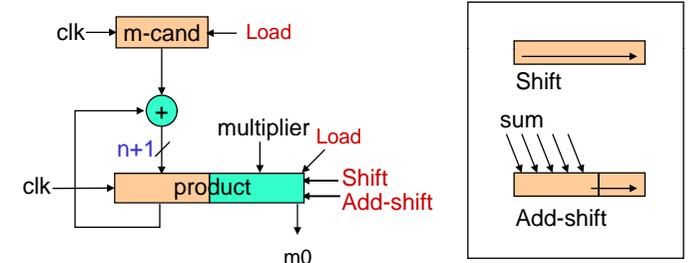
- 13 x 11 = 143



# 8.5 빠른 곱셈기

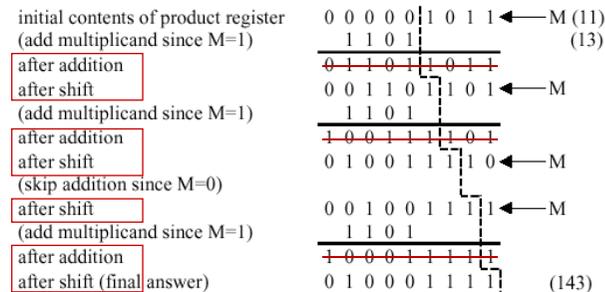
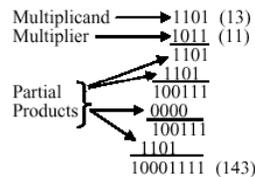
## datapath

- multiplier를 product의 하위부분을 공유 사용
- add 결과 저장과 shift를 한 번에 수행 → 동작: shift 또는 add-shift



# 빠른 곱셈기

## 곱셈 수행 과정



dividing line between product and multiplier

# 빠른 곱셈기

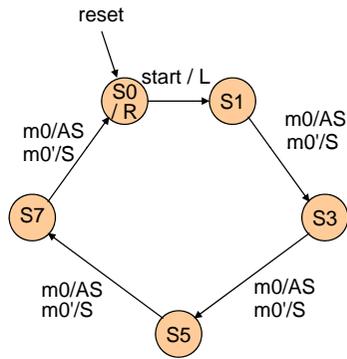
```

module Datapath (product, m0, word1, word2, Load, Shift, Add_Shift, reset, clock);
    output [7:0] product;
    output m0;
    input [3:0] word1, word2;
    input Load, Shift, Add_Shift, clock, reset;
    reg [7:0] product;
    reg [3:0] multiplier, multiplicand;
    wire [4:0] sum; // 5-bit

    assign m0 = product[0];
    assign sum = product[7:4] + multiplicand;
    always @ (posedge clock or posedge reset) begin
        if (reset) begin multiplicand <= 0; product <= 0; end
        else if (Load) begin
            multiplicand <= word1;
            product <= {4'b0, word2};
        end else if (Shift) begin
            product <= product >> 1; // product <= {1'b0, product[7:1]};
        end else if (Add_Shift) begin
            product[7:3] <= sum; // product <= {sum, product[3:1]};
            product[2:0] <= product[3:1];
        end
    end
endmodule
    
```

## 빠른 곱셈기

### controller의 상태도



L: Load  
R: Ready  
S: Shift  
AS: Add-Shift

## 빠른 곱셈기

```

module Controller (Load, Shift, Add_Shift, Ready, m0, Start, reset, clock, state);
    output Load, Shift, Add_Shift, Ready;
    input m0, Start, clock, reset;
    output [2:0] state;
    reg [2:0] state, next_state;
    parameter S0 = 0, S1 = 4, S3 = 5, S5 = 6, S7 = 7;
    wire Load, Shift, Add_Shift, Ready;
    
```

```

always @ (posedge clock or posedge reset) // State transitions
    if (reset) state <= S0;
    else state <- next_state;
    
```

(계속)

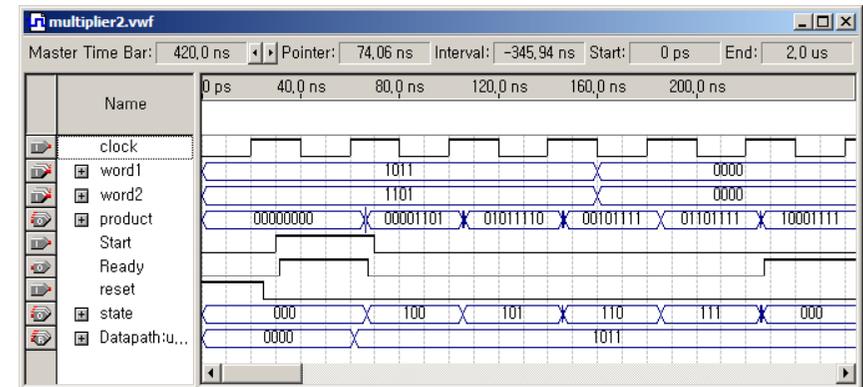
## 빠른 곱셈기

```

always @ (state or Start or m0) begin // Next state logic
    case (state)
        S0: if (Start) next_state = S1; else next_state = S0;
        S1: next_state = S3;
        S3: next_state = S5;
        S5: next_state = S7;
        S7: next_state = S0;
        default: next_state = S0;
    endcase
end
// outputs
assign Add_Shift = (state==S1 || state==S3 || state==S5 || state==S7) & m0;
assign Shift = (state==S1 || state==S3 || state==S5 || state==S7) & ~m0;
assign Load = (state==S0) && Start;
assign Ready = (state==S0) && ~reset;
endmodule
    
```

## 빠른 곱셈기

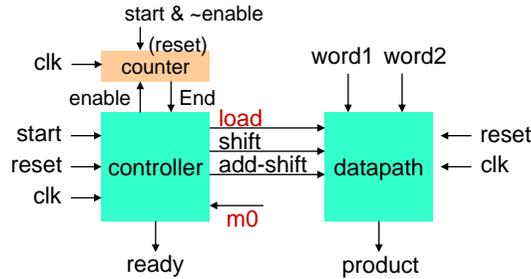
### Simulation



## 8.6 카운터를 사용한 곱셈기 제어

### 곱셈기 제어의 문제점 및 해결책

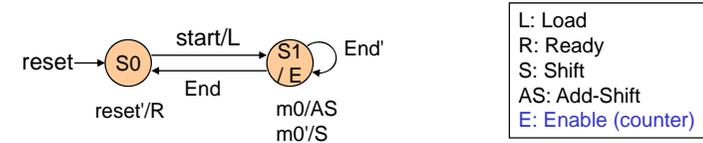
- 문제점: 비트 수를 변경하면 제어기를 수정해야 함
- 해결책: 카운터를 사용하여 제어기 설계



**counter 동작**  
 곱셈 시작할 때에 count < 0  
 각 곱셈 단계에서 count 증가  
 counter 출력이 N-1이면 End 신호 출력

## 카운터를 사용한 곱셈기 제어

### 상태도



L: Load  
 R: Ready  
 S: Shift  
 AS: Add-Shift  
 E: Enable (counter)

### Verilog description

```

    counter
    module counter(count, Enable, reset, clock);
        output [3:0] count;
        input Enable, reset, clock;
        reg [3:0] count;

        always @(posedge clock)
            if (reset) count <= 0;
            else if (Enable) count <= count + 1;
    endmodule
    
```

## 카운터를 사용한 곱셈기 제어

```

    module multiplier3 (product, Ready, word1, word2, Start, reset,
        clock);
        parameter N=4;
        output [2*N-1:0] product;
        output Ready;
        input [N-1:0] word1, word2;
        input Start, clock, reset;
        wire m0, Load, Shift, Add_Shift, Enable, End;
        wire [3:0] count;

        Datapath u1 (product, m0, word1, word2, Load, Shift, Add_Shift,
            reset, clock);
        Controller u2 (Load, Shift, Add_Shift, Enable, Ready, m0, Start,
            End, reset, clock);
        counter u3 (count, Enable, Start & ~Enable, clock);
        assign End = (count==N-1);
    endmodule
    
```

counter reset 신호

datapath는 앞의 빠른 곱셈기와 같음

## 카운터를 사용한 곱셈기 제어

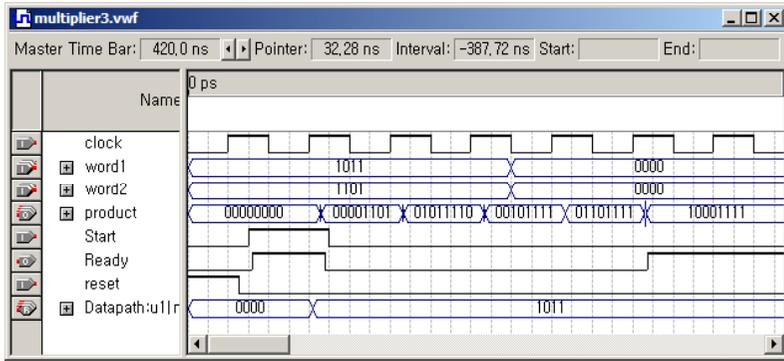
```

    module Controller (Load, Shift, Add_Shift, Enable, Ready, m0, Start,
        End, reset, clock);
        output Load, Shift, Add_Shift, Enable, Ready;
        input m0, Start, End, clock, reset;
        reg state, next_state;
        parameter S0 = 0, S1 = 1;
        wire Shift, Add_Shift, Load, Ready;

        always @(posedge clock or posedge reset) // State transitions
            if (reset) state <= S0; else state <= next_state;
        always @(state or Start or m0) // Next state and control logic
            case (state)
                S0: if (Start) next_state = S1; else next_state = S0;
                S1: if (End) next_state = S0; else next_state = S1;
            endcase
        assign Add_Shift = (state==S1) & m0;
        assign Shift = (state==S1) & ~m0;
        assign Load = (state==S0) & Start;
        assign Enable = (state==S1);
        assign Ready = (state==S0) && ~reset;
    endmodule
    
```

# 카운터를 사용한 곱셈기 제어

## Simulation (N=4)



# 8.7 signed 곱셈기

## Binary notation: $B = b_{n-1}b_{n-2} \dots b_1b_0$

- unsigned(B) =  $\sum_{i=0}^{n-1} b_i 2^i$
- signed(B) =  $-b_{n-1} 2^{n-1} + \sum_{i=0}^{n-2} b_i 2^i$

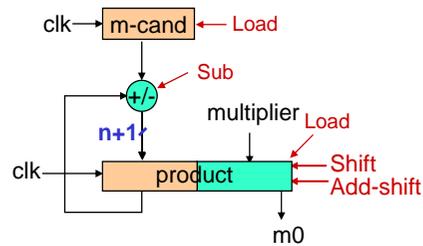
## unsigned 곱셈기를 signed 곱셈기로 간단하게 수정하는 방법

- 최상위 bit가 1이면 multiplicand를 더하는 것 대신에 뺄셈을 수행
- 뺄셈을 수행할 수 있도록 datapath를 수정 → 2의 보수 덧셈 (가감산기)
- adder를 signed number에 대한 덧셈/뺄셈에 사용할 때에는 carry-out은 사용하지 않음 (대신에 n+1비트 가감산수행)
- product를 shift-right 시킬 때에 **보호(최상위 bit)를 그대로 유지**해야 함

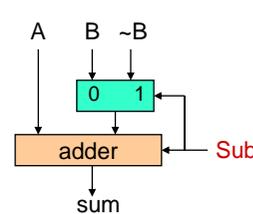
# signed 곱셈기

## datapath

- n+1 비트 가감산기 사용
  - multiplicand는 n+1비트로 부호확장
- carry출력은 사용하지 않음

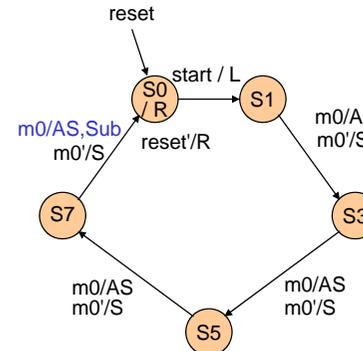


## adder/subtractor



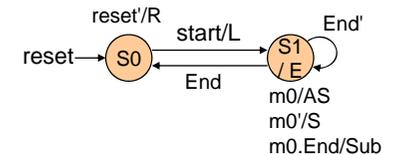
# signed 곱셈기

## controller의 상태도



L: Load  
R: Ready  
S: Shift  
AS: Add-Shift  
Sub: Sub

## 카운터 사용한 상태도



## signed 곱셈기

```

module Datapath (product, m0, word1, word2, Load, Shift, Add_Shift, Sub,
                reset, clock);
    output [7:0] product;
    output m0;
    input [3:0] word1, word2;
    input Load, Shift, Add_Shift, Sub, clock, reset;
    reg [8:0] eproduct; // 9-bit
    reg [4:0] multiplicand; // 5-bit
    wire [4:0] sum; // 5-bit

    assign product = eproduct[7:0];
    assign m0 = eproduct[0];
    always @ (posedge clock or posedge reset) begin
        if (reset) begin multiplicand <= 0; eproduct <= 0; end
        else if (Load) begin
            multiplicand <= {word1[3],word1}; // sign-extension
            eproduct <= {5'b0, word2};
        end else if (Shift) // signed shift-right
            eproduct <= {eproduct[8], eproduct[8:1]};
        else if (Add_Shift) // signed add & shift right
            eproduct <= {sum[4], sum, eproduct[3:1]};
        end
        assign sum = eproduct[8:4] + (Sub ? ~multiplicand : multiplicand) + Sub;
    end
endmodule

```

→ adder/subtractor (sub=1일때 뺄셈)

## signed 곱셈기

```

module Controller (Load, Shift, Add_Shift, Sub, Enable, Ready, m0, Start,
                  End, reset, clock);
    output Load, Shift, Add_Shift, Sub, Enable, Ready;
    input m0, Start, End, clock, reset;
    reg state, next_state;
    parameter S0 = 0, S1 = 1;

    always @ (posedge clock or posedge reset) // State transitions
        if (reset) state <= S0; else state <= next_state;

    always @ (state or Start or m0) // Next state logic
        case (state)
            S0: if (Start) next_state = S1; else next_state = S0;
            S1: if (End) next_state = S0; else next_state = S1;
        endcase

    // control output logic
    assign Add_Shift = (state==S1) & m0;
    assign Sub = (state==S1) & End & m0;
    assign Shift = (state==S1) & ~m0;
    assign Load = (state==S0) && Start;
    assign Enable = (state==S1);
    assign Ready = (state==S0) && ~reset;
endmodule

```

## signed 곱셈기

```

module smultiplier(product, Ready, word1, word2, Start, clock, reset);
    parameter N=4;
    output [2*N-1:0] product;
    output Ready;
    input [N-1:0] word1, word2;
    input Start, clock, reset;
    wire m0, Load, Shift, Add_Shift, Sub, Enable, End;
    wire [3:0] count;

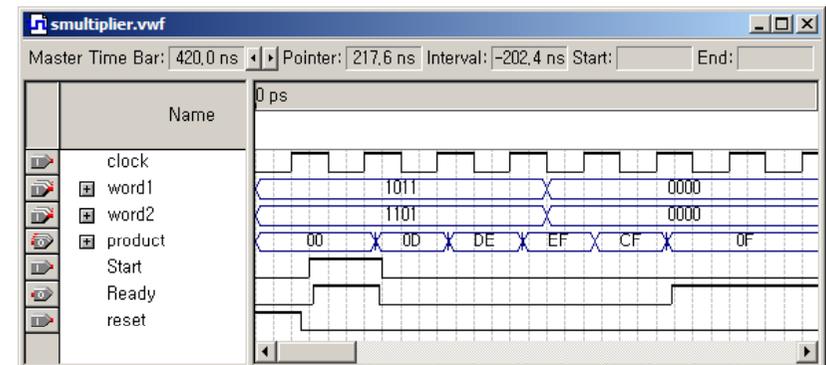
    Datapath u1 (product, m0, word1, word2, Load, Shift, Add_Shift, Sub,
                reset, clock);
    Controller u2 (Load, Shift, Add_Shift, Sub, Enable, Ready, m0, Start,
                 End, reset, clock);
    counter u3 (count, Enable, Start & ~Enable, clock);
    assign End = (count==N-1);
endmodule

```

counter는 앞의 곱셈기와 같음

## signed 곱셈기: Verilog description

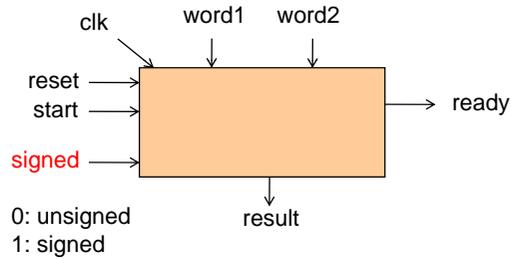
- simulation
  - ( 5 ) \* ( 3 ) = 15



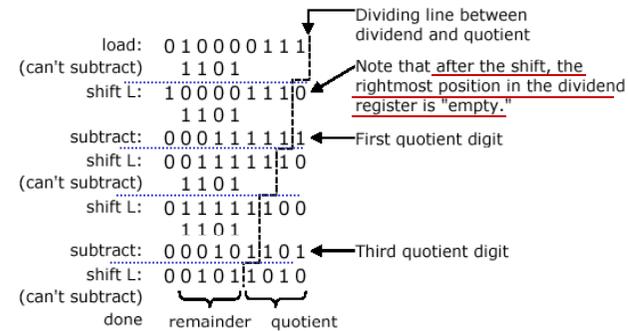
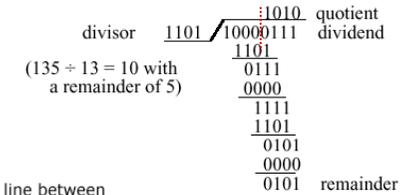
# 통합 곱셈기

## 통합 곱셈기

- 제어신호에 따라서 unsigned 곱셈 또는 signed 곱셈 수행
- 두 가지 곱셈이 모두 가능하도록 datapath와 controller를 수정



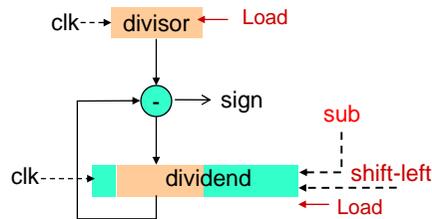
# 순차 나눗셈 - unsigned



# 순차 나눗셈기 - unsigned

## datapath

- dividend와 remainder/quotient를 함께 사용
  - 나눗셈 후에 dividend는 remainder/quotient로 대체됨
- divisor를 shift-right하는 것 대신에 dividend를 shift-left함
- n+1 비트 subtractor 사용



# 나눗셈 알고리즘

```

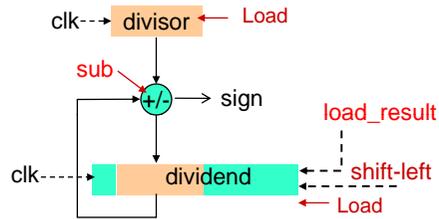
Dividend (9-bit) ← word1 (8-bit), Divisor ← word1 (4-bit)
repeat 4 times do
  Shift Left Dividend
  Diff = Dividend[8:4] - {0, Divisor[3:0]} // 5-bit subtraction
  if (Diff < 0) // Diff[4]=1
    Dividend[0] ← 0 // Quotient = 0
  else // Diff[4]=0
    Dividend[0] ← 1, Dividend[8:4] ← Diff // Quotient = 1
end
Remainder ← Dividend[7:4], Quotient ← Dividend[3:0]

if (Dividend[8:4] >= Divisor) then Overflow
    
```

## 순차 나눗셈기 - signed

### ■ datapath

- n+1 비트 adder/subtractor 사용



## 나눗셈 알고리즘

Dividend (8-bit)  $\leftarrow$  word1 (8-bit), Divisor  $\leftarrow$  word1 (4-bit)

**repeat** 4 times **do**

- Shift Left Dividend

- if (Dividend and Divisor have same sign)

- Result = Dividend[8:4] - {Divisor[3], Divisor[3:0]}

- else

- Result = Dividend[8:4] + {Divisor[3], Divisor[3:0]}

- if (Dividend and Result have same sign or Result is 0)

- Dividend[0]  $\leftarrow$  1, Dividend[8:4]  $\leftarrow$  Result // Quotient = 1

- else

- Dividend[0]  $\leftarrow$  0

- // Quotient = 0

**end**

Remainder  $\leftarrow$  Dividend[7:4]

if (word1 and word2 have same sign) Quotient  $\leftarrow$  Dividend[3:0]

else Quotient  $\leftarrow$  2's complement of Dividend[3:0]